

Space Vector PWM

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Space vectors

- The origin of space vectors lies in rotating mmf in machines.
- The resultant mmf for a three phase system is a rotating mmf having a fixed magnitude and direction at every instant of time.
- Space vector is a mathematical concept which is useful for visualizing the effect of three phase variables in space.



Space vectors

- Resultant space vector for load phase voltage or current are defined as,

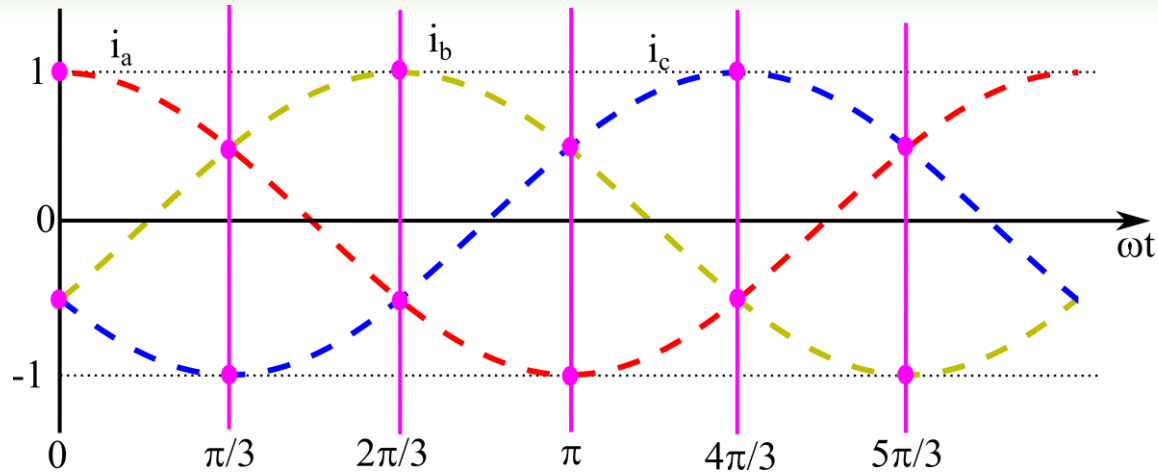
- $$\mathbf{V}_R(\mathbf{t}) = \frac{2}{3} \left[v_{An}(t) + v_{Bn}(t)e^{\frac{j2\pi}{3}} + v_{Cn}(t)e^{\frac{j4\pi}{3}} \right]$$

- $$\mathbf{I}_R(\mathbf{t}) = \frac{2}{3} \left[i_A(t) + i_B(t)e^{\frac{j2\pi}{3}} + i_C(t)e^{\frac{j4\pi}{3}} \right]$$

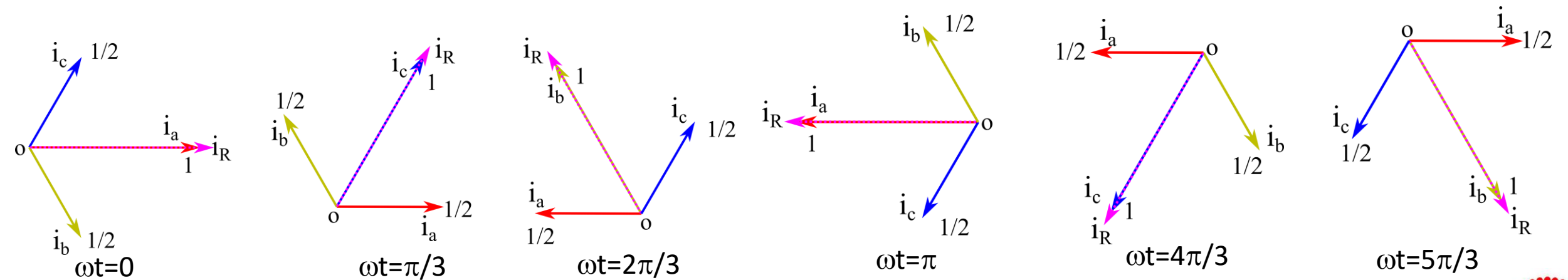
- The space vectors $\mathbf{V}_R(\mathbf{t})$ or $\mathbf{I}_R(\mathbf{t})$ have both magnitude and angle. Individual voltages/currents can be balanced or unbalanced and need not be sinusoidal.



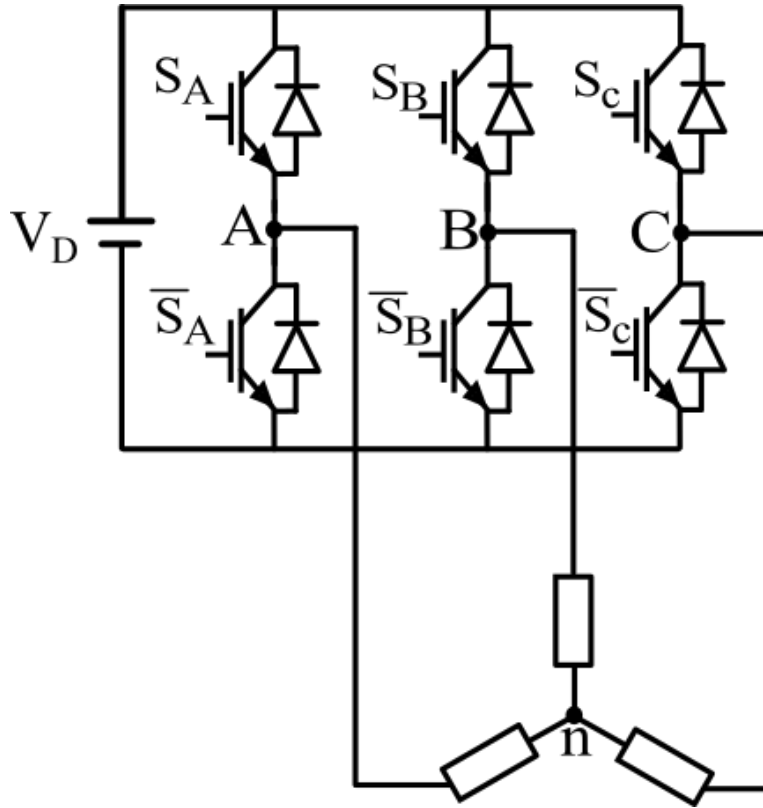
Current space Vector



- For the sinusoidal three phase currents, the resultant current space vector is shown.
- The resultant space vector (pink) is rotating at a uniform speed and having a constant radius.



Space vectors



- The pole voltage of one phase of the converter has two switching states: 1 ($=V_D$) and 0 ($=0$).
- The converter has total eight switching states ($2*2*2=8$). These are: (000,111,100,110,010,011,001,101).
- There are six active vectors and two zero vectors.
- What is the load phase voltage space vector for 100 combination?



Space vector for 100 combination

- $v_{AO}(t) = VD$, $v_{BO}(t) = 0$, $v_{CO}(t) = 0$
- $v_{An}(t) = \frac{2}{3}v_{AO}(t) - \frac{1}{3}v_{BO}(t) - \frac{1}{3}v_{CO}(t) = \frac{2}{3}V_D$
- $v_{Bn}(t) = \frac{2}{3}v_{BO}(t) - \frac{1}{3}v_{CO}(t) - \frac{1}{3}v_{AO}(t) = -\frac{1}{3}V_D$
- $v_{Cn}(t) = \frac{2}{3}v_{CO}(t) - \frac{1}{3}v_{AO}(t) - \frac{1}{3}v_{BO}(t) = -\frac{1}{3}V_D$
- $\mathbf{V}_R(\mathbf{t}) = \frac{2}{3} \left[v_{An}(t) + v_{Bn}(t)e^{\frac{j2\pi}{3}} + v_{Cn}(t)e^{\frac{j4\pi}{3}} \right] = \frac{2}{3}V_D e^{j0}$
- Similarly we can deduce the resultant space vector for other combinations.



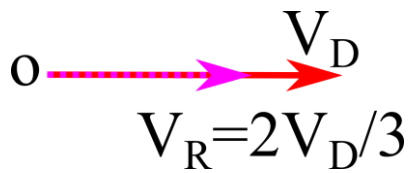
Space vector for all combinations

Space Vector	Switching States	Resultant space vector ($\vec{V}_R(t)$)	
V0	000	$\vec{V}_0 = 0$	Zero Vector
V1	100	$\vec{V}_1 = \frac{2}{3}V_D e^{j0}$	Active Vector
V2	110	$\vec{V}_2 = \frac{2}{3}V_D e^{j\pi/3}$	
V3	010	$\vec{V}_3 = \frac{2}{3}V_D e^{j2\pi/3}$	
V4	011	$\vec{V}_4 = \frac{2}{3}V_D e^{j3\pi/3}$	
V5	001	$\vec{V}_5 = \frac{2}{3}V_D e^{j4\pi/3}$	
V6	101	$\vec{V}_6 = \frac{2}{3}V_D e^{j5\pi/3}$	
V7	111	$\vec{V}_7 = 0$	Zero Vector

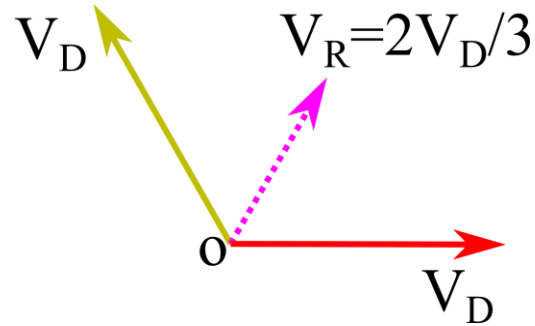


Graphical way

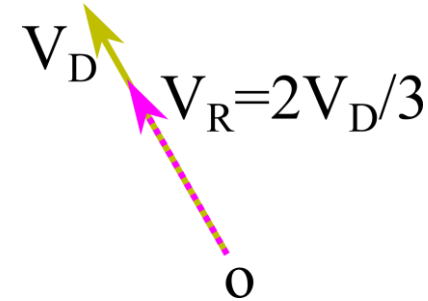
100 Condition



110 Condition

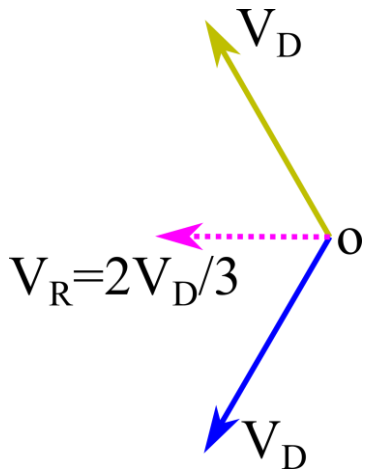


010 Condition

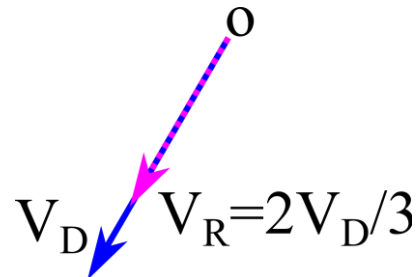


- The space vectors can be obtained also from a graphical method.

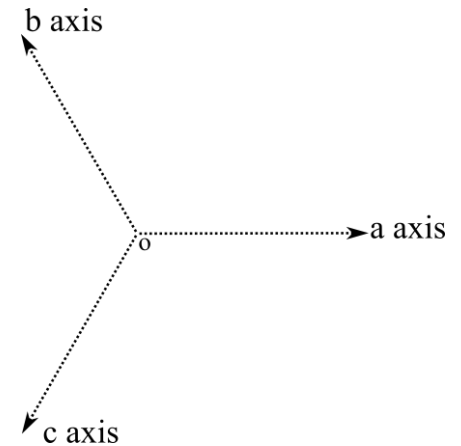
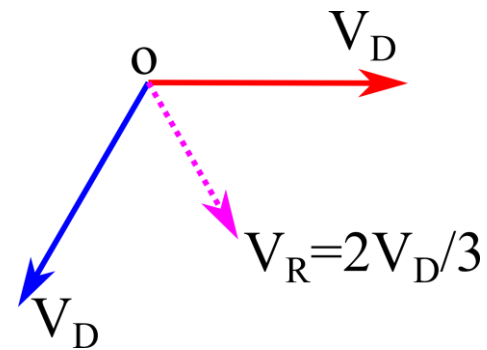
011 Condition



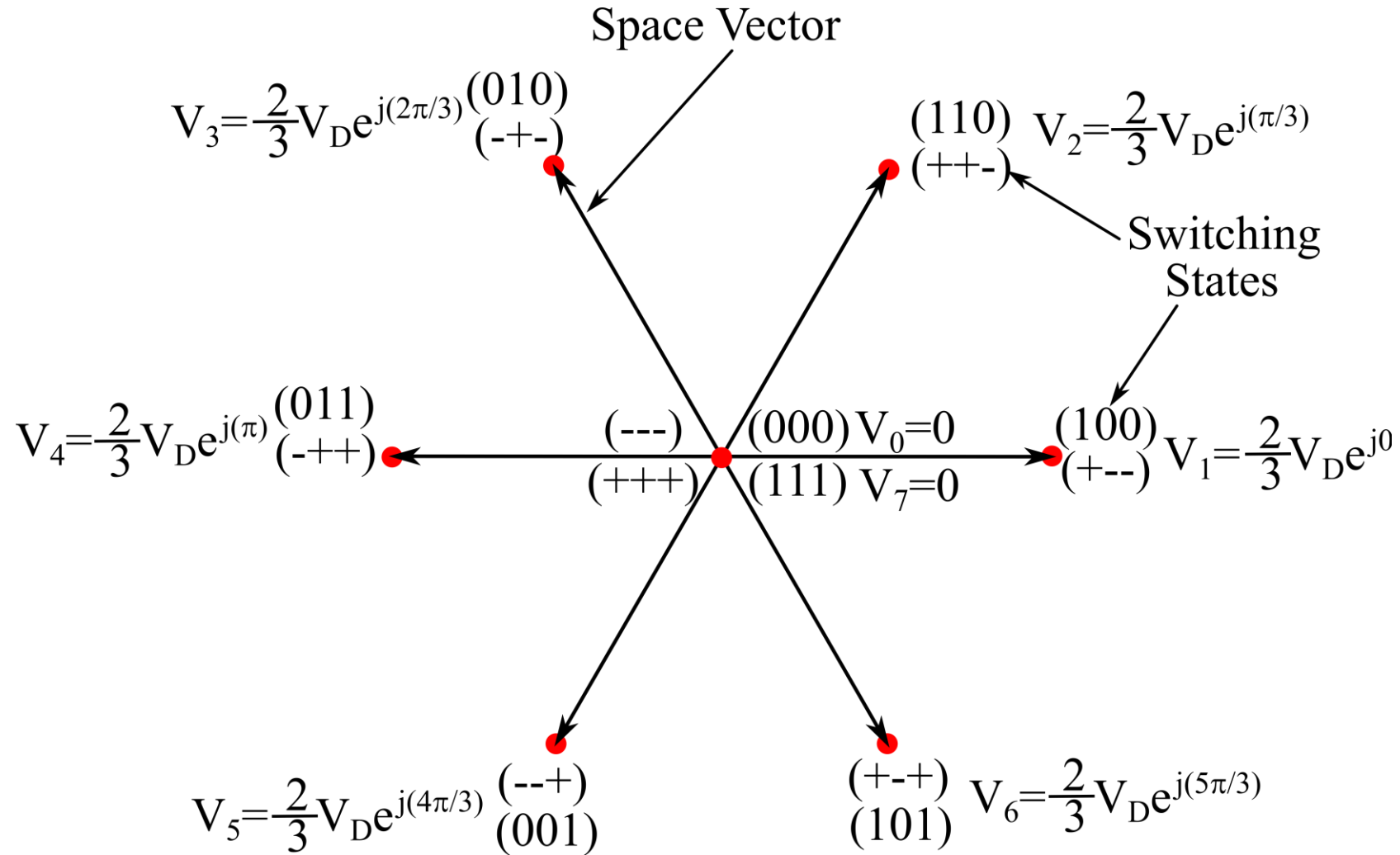
001 Condition



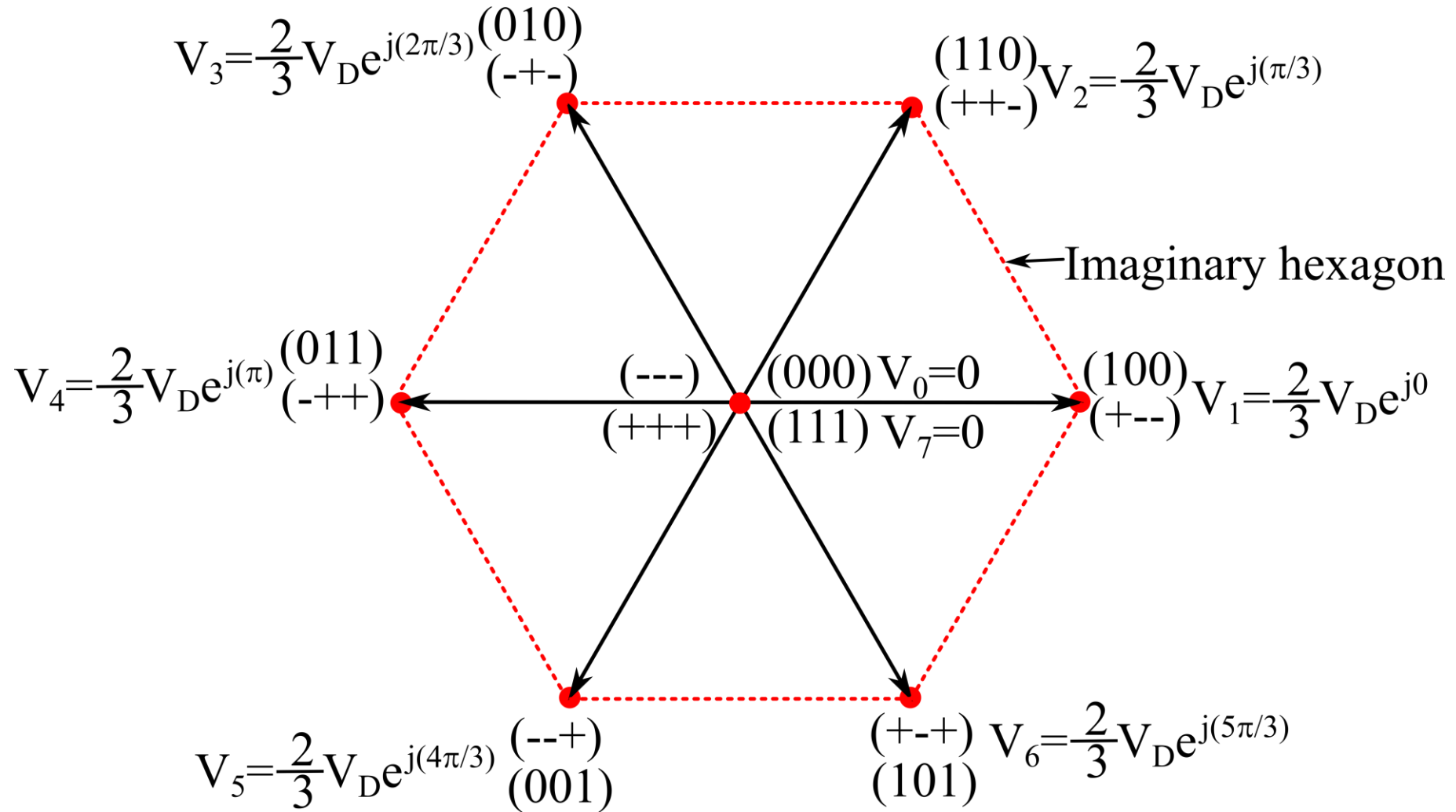
101 Condition



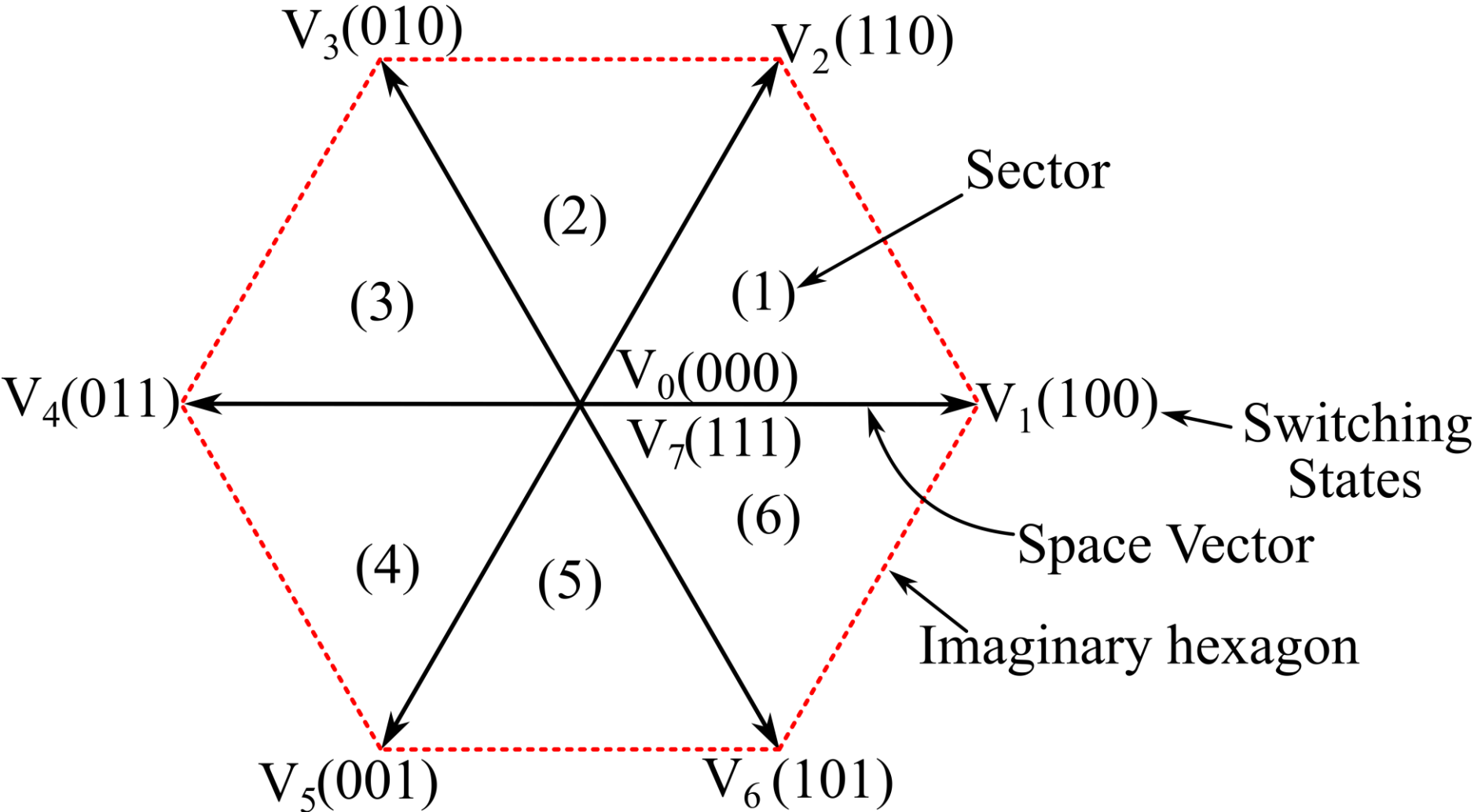
Eight space vectors



Boundary of space vector diagram



Sectors in space vector diagram



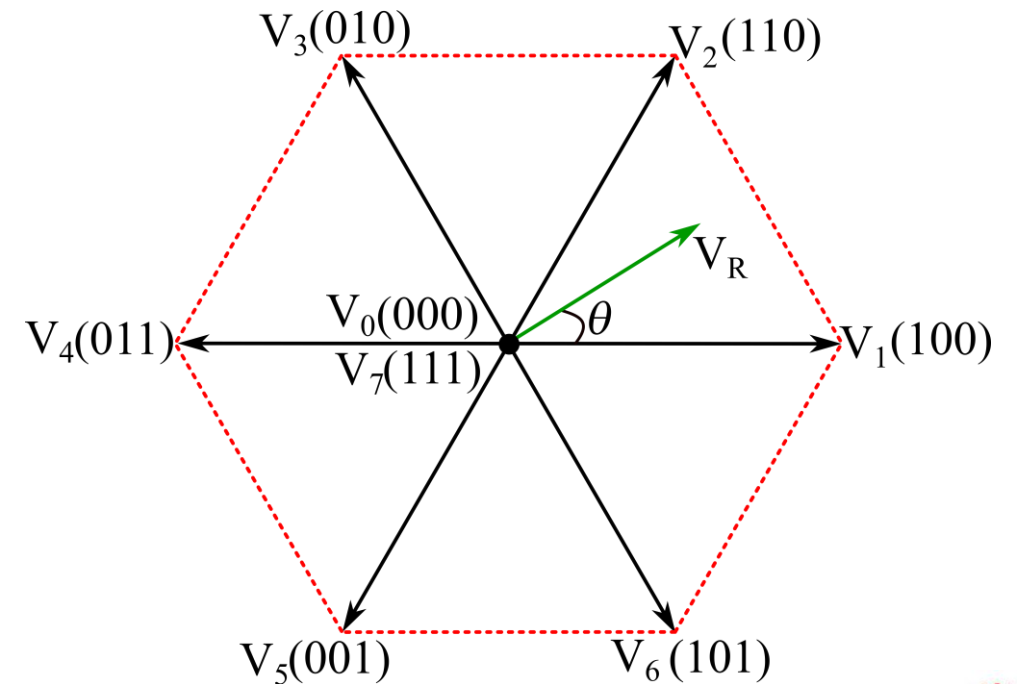
Space Vector PWM

- How to switch the eight vectors so that the correct voltage is impressed on the load?
- Space vector PWM is an extension of sine triangle PWM. Here the PWM is done by using space vectors.

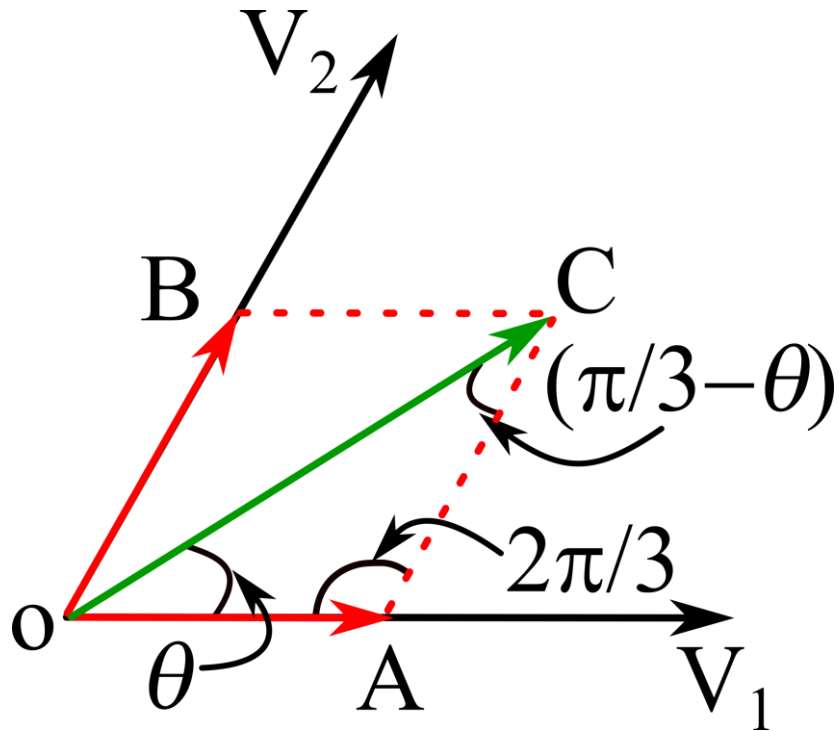


Space Vector PWM

- The space vectors are switched for certain duration of time in a cycle so as to produce the resultant vector.
- $V_R T_S = V_1 T_1 + V_2 T_2 + V_0 T_0 = V_1 T_1 + V_2 T_2 + V_0 T_{01} + V_0 T_{07}$
- $T_S = T_1 + T_2 + T_0$
- In space vector PWM, $T_{01} = T_{07} = T_0/2$



Mathematical expression of timings



- $\frac{OA}{\sin(\frac{\pi}{3}-\theta)} = \frac{OB}{\sin(\theta)} = \frac{OC}{\sin(\frac{2\pi}{3})}$
- $\frac{V_1 T_1}{\sin(\frac{\pi}{3}-\theta)} = \frac{V_2 T_2}{\sin(\theta)} = \frac{V_R T_S}{\sin(\frac{2\pi}{3})}$
- $T_1 = \sin(\frac{\pi}{3} - \theta) \frac{V_R}{V_1} \frac{2}{\sqrt{3}} T_S = \sin(\frac{\pi}{3} - \theta) \frac{V_R}{V_D} \sqrt{3} T_S$
- $T_2 = \sin \theta \frac{V_R}{V_1} \frac{2}{\sqrt{3}} T_S = \sin \theta \frac{V_R}{V_D} \sqrt{3} T_S$
- $T_0 = T_S - T_1 - T_2$
- What happens at $\theta = 0$, and $V_R = 2/3 V_D$?



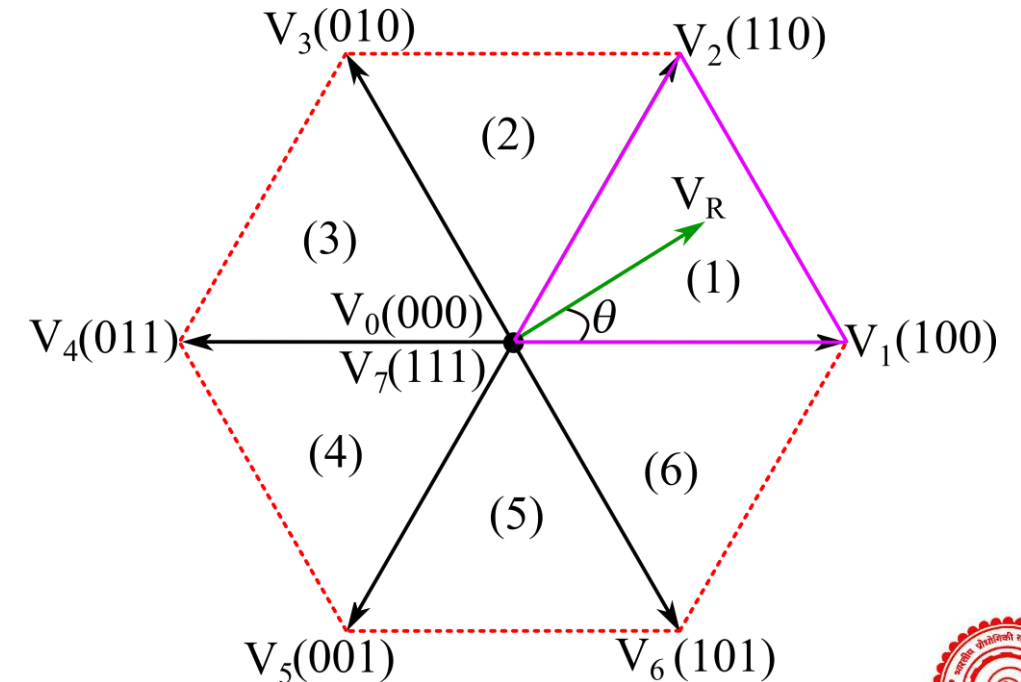
Zero vector

- Usually the zero vectors are kept equal. This gives the best harmonic performance.
- $T_0 = T_S - T_1 - T_2$ is divided into equal parts of $T_0/2$ at the beginning and end of the cycle i.e. $T_{01} = T_{07} = \frac{T_0}{2}$
- For special switching sequences (e.g. discontinuous PWM), the division is made not equal.



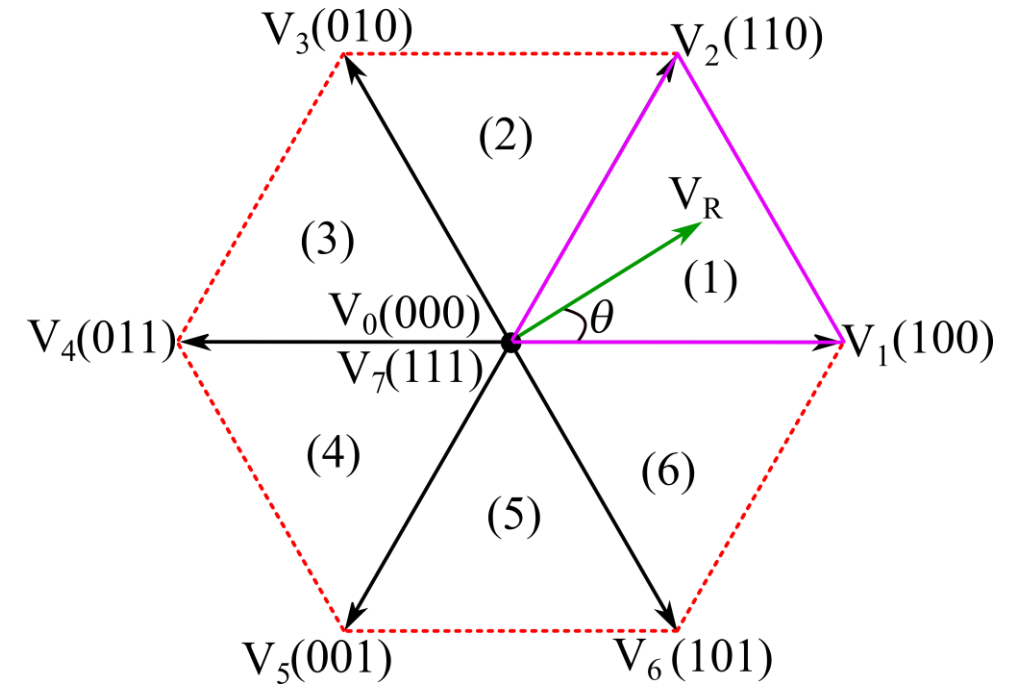
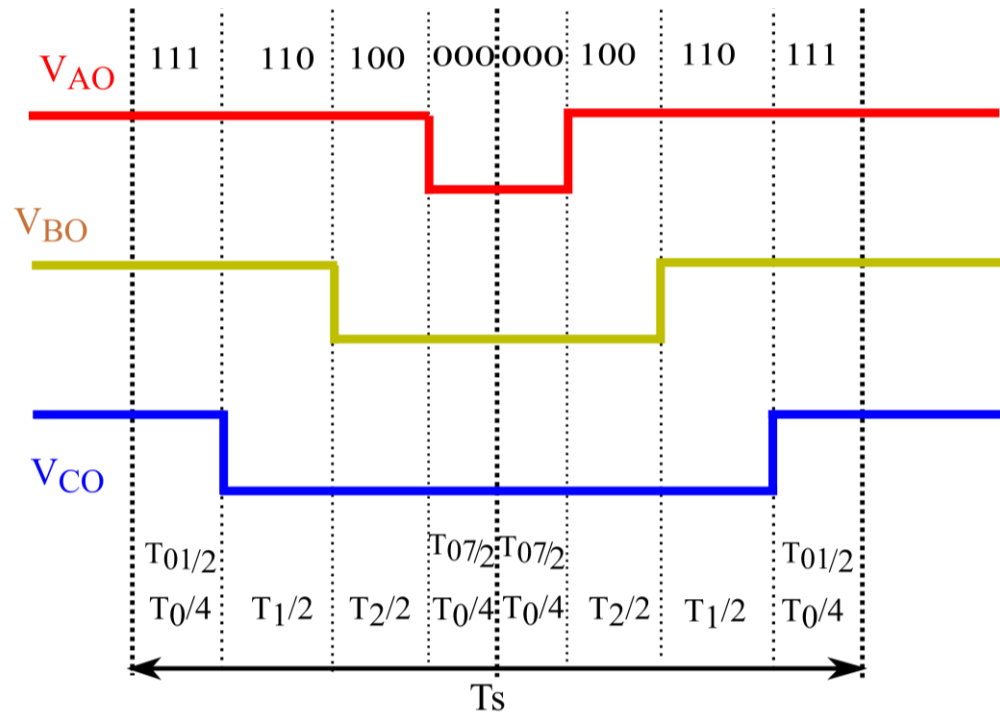
Example of switching

- For example, we can switch in a switching cycle T_s : 111 for T_{01} time period, 110 for T_1 time period, 100 for T_2 time period and 000 for T_{07} time period. This will realize the reference vector (V_R) in the switching cycle T_s .
- $V_R T_S = V_1 T_1 + V_2 T_2 + V_0 T_0 = V_1 T_1 + V_2 T_2 + V_0 T_{01} + V_0 T_{07}$



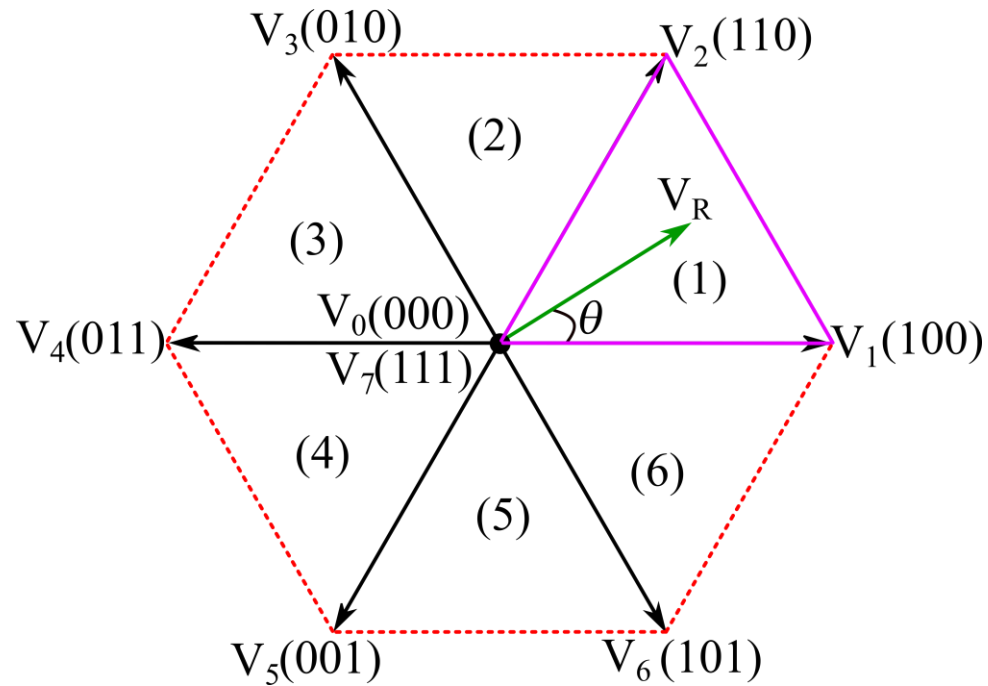
Example of switching

- The instantaneous pole voltages can be seen from the diagram. The switching sequence is 111-110-100-000-100-110-111 and so on in sector 1.
- The sequence ensures minimum switching.

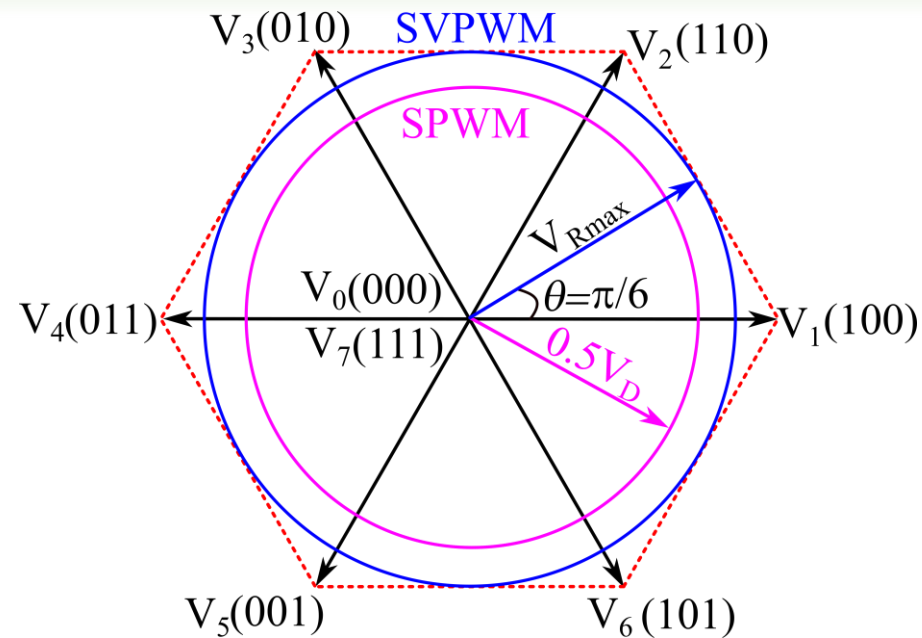


Example of switching

- Similarly, 111-110-010-000-010-110-111 and so on in sector 2.



What is the maximum voltage?



- The maximum voltage is obtained in linear modulation when the inscribed circle touches the hexagon.
- $V_{Rmax} = \frac{2}{3} V_D \cos \frac{\pi}{6} = 0.577 V_D$
- In sine-PWM the peak AC voltage that was obtained was $0.5 V_D$.

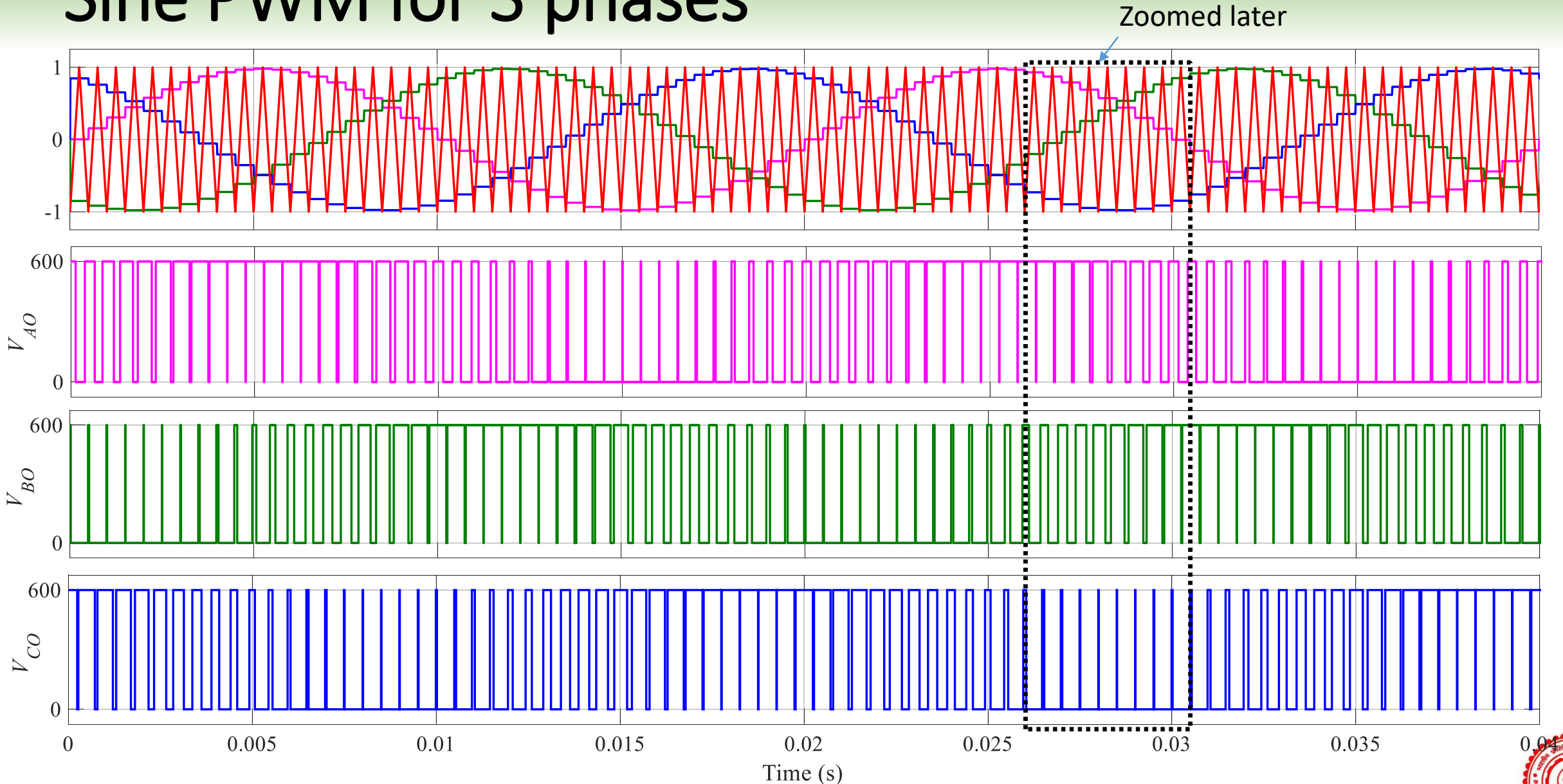


How to realize using carriers?

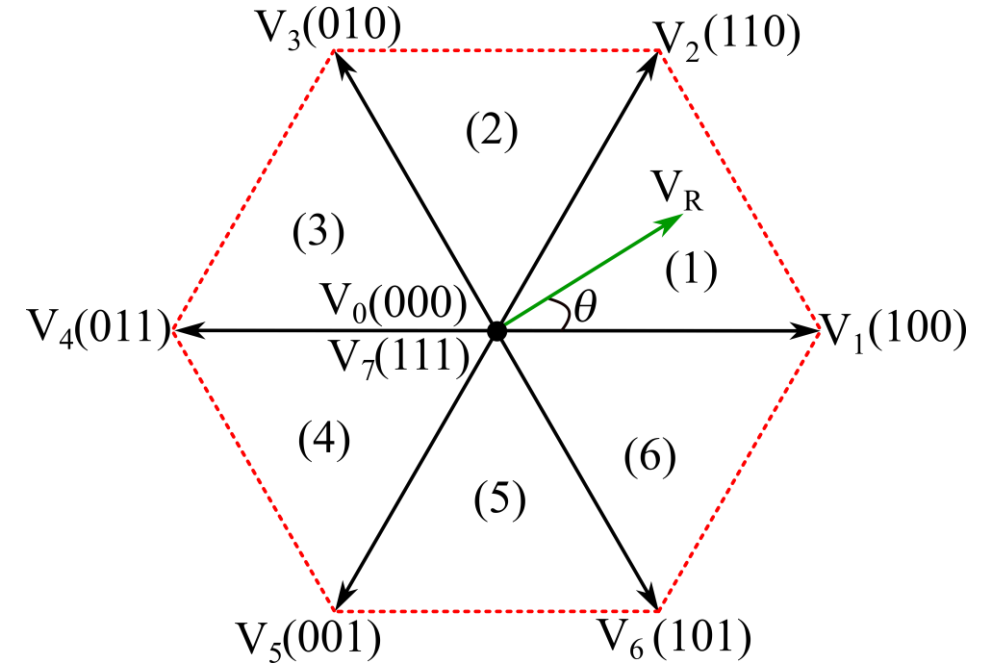
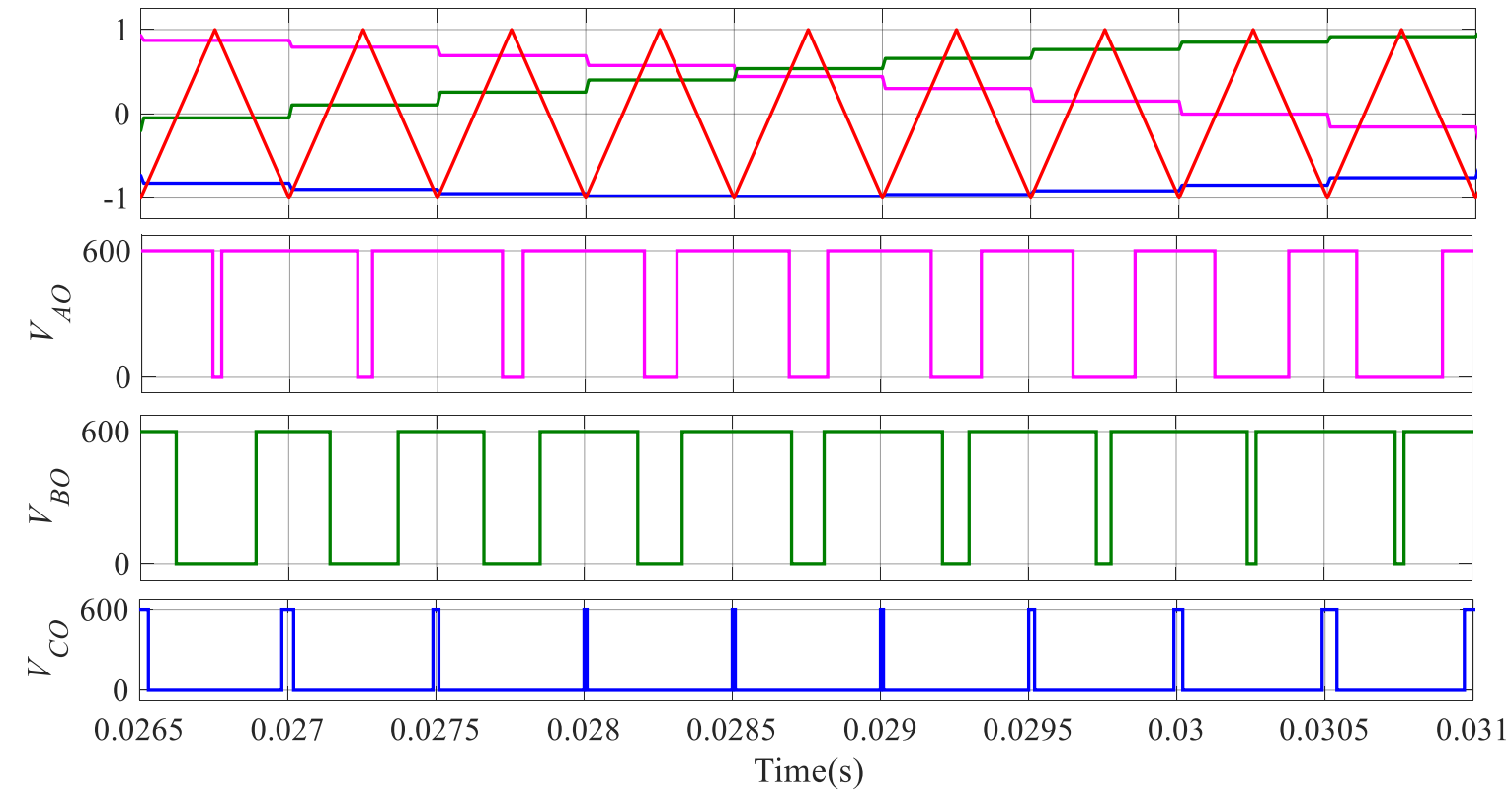
- The SVPWM technique discussed so far involves substantial calculation, sector identification etc.
- It can be done very easily using carriers where no calculation, sector identification or switching sequence design is required.
- In order to realize SVPWM through carriers, we can observe the sine PWM more in details.



Sine PWM for 3 phases



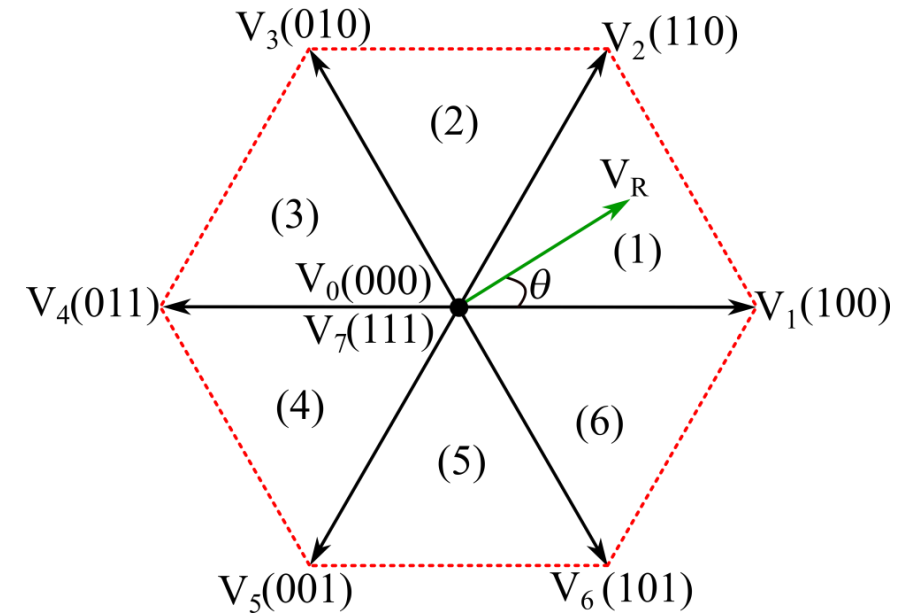
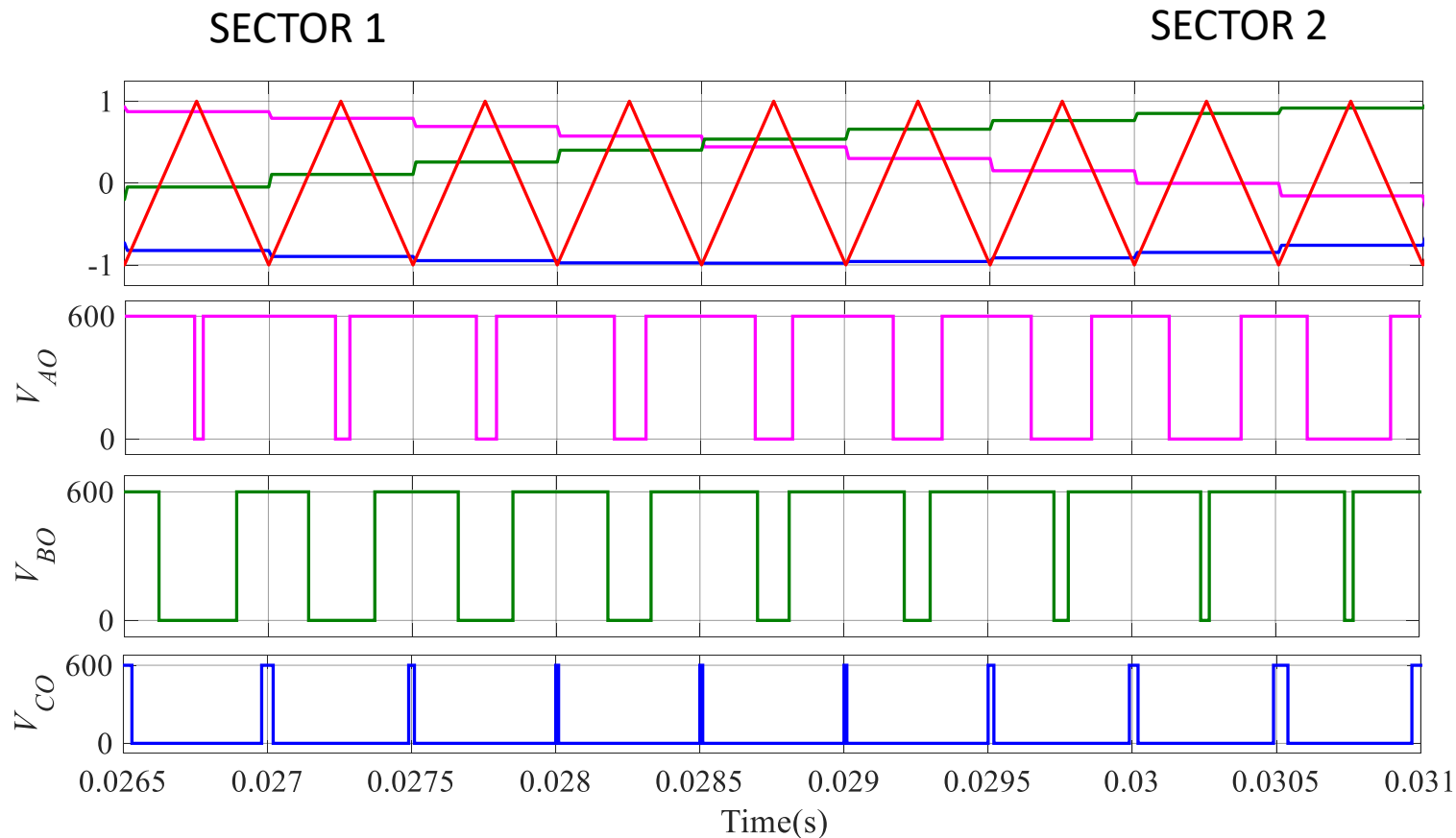
Zoomed view in sine PWM



- What is the pattern of switching?



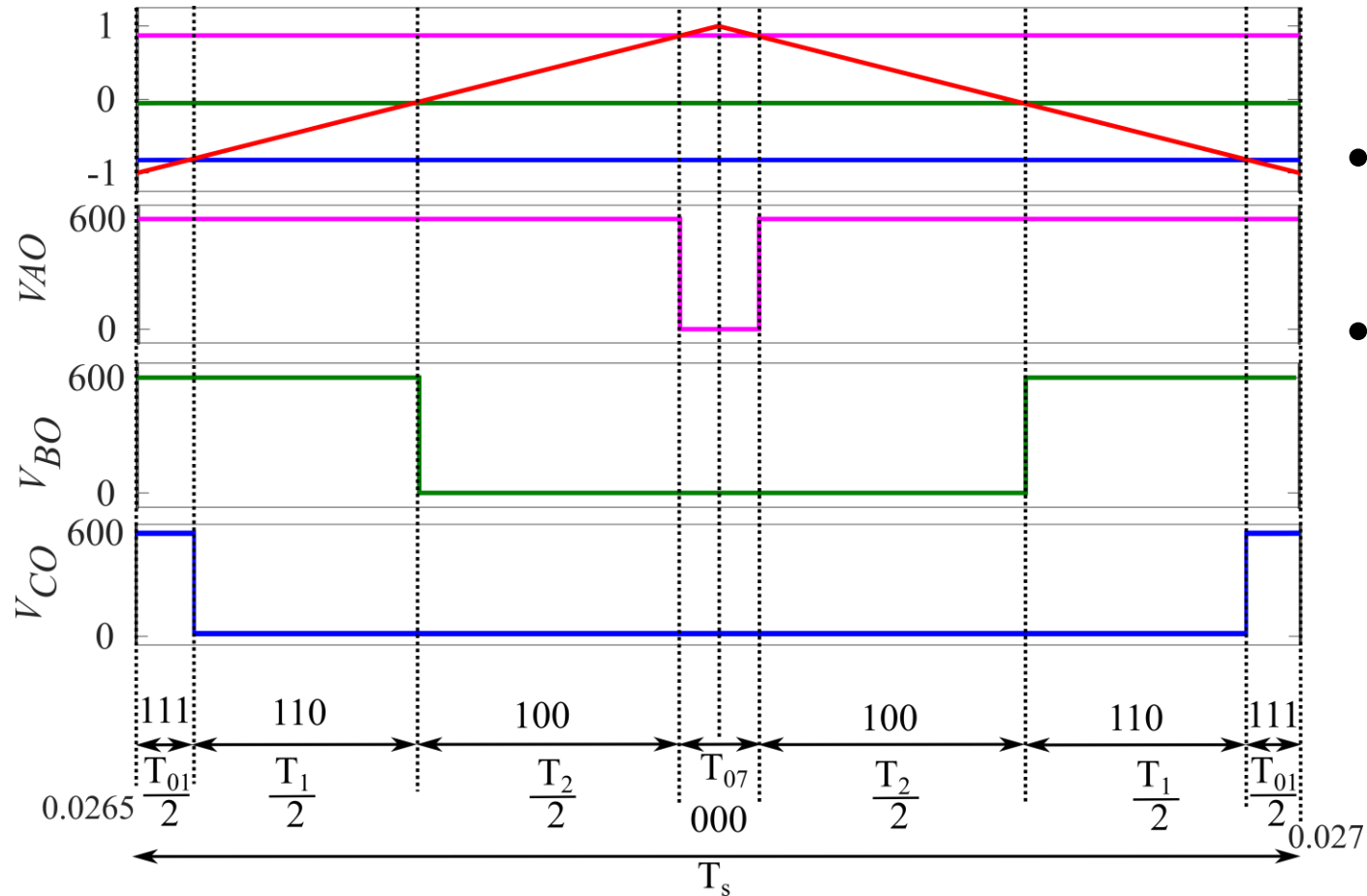
Zoomed view in sine PWM



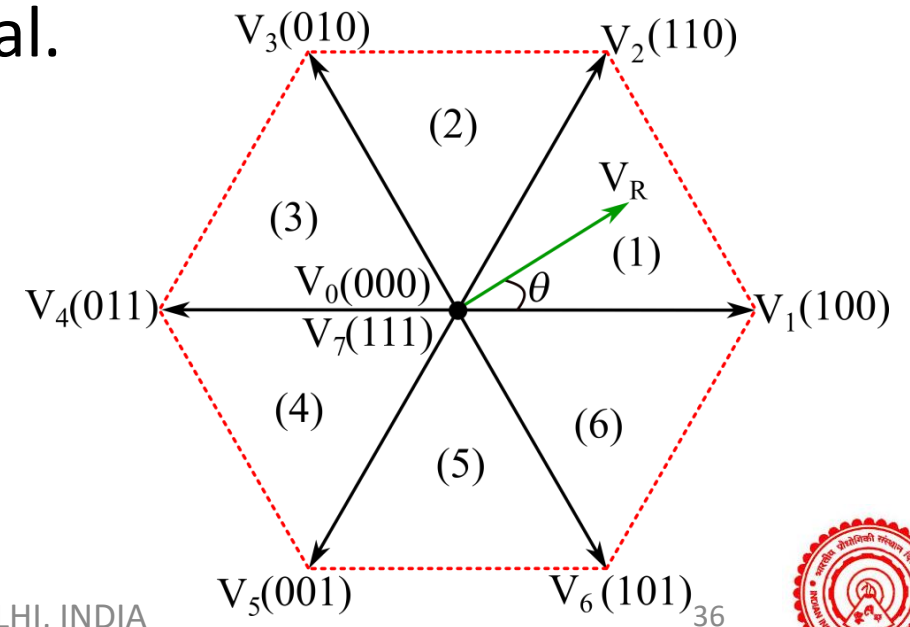
- 111-110-100-000-100-110-111 and so on in sector 1.
- 111-110-010-000-010-110-111 and so on in sector 2.



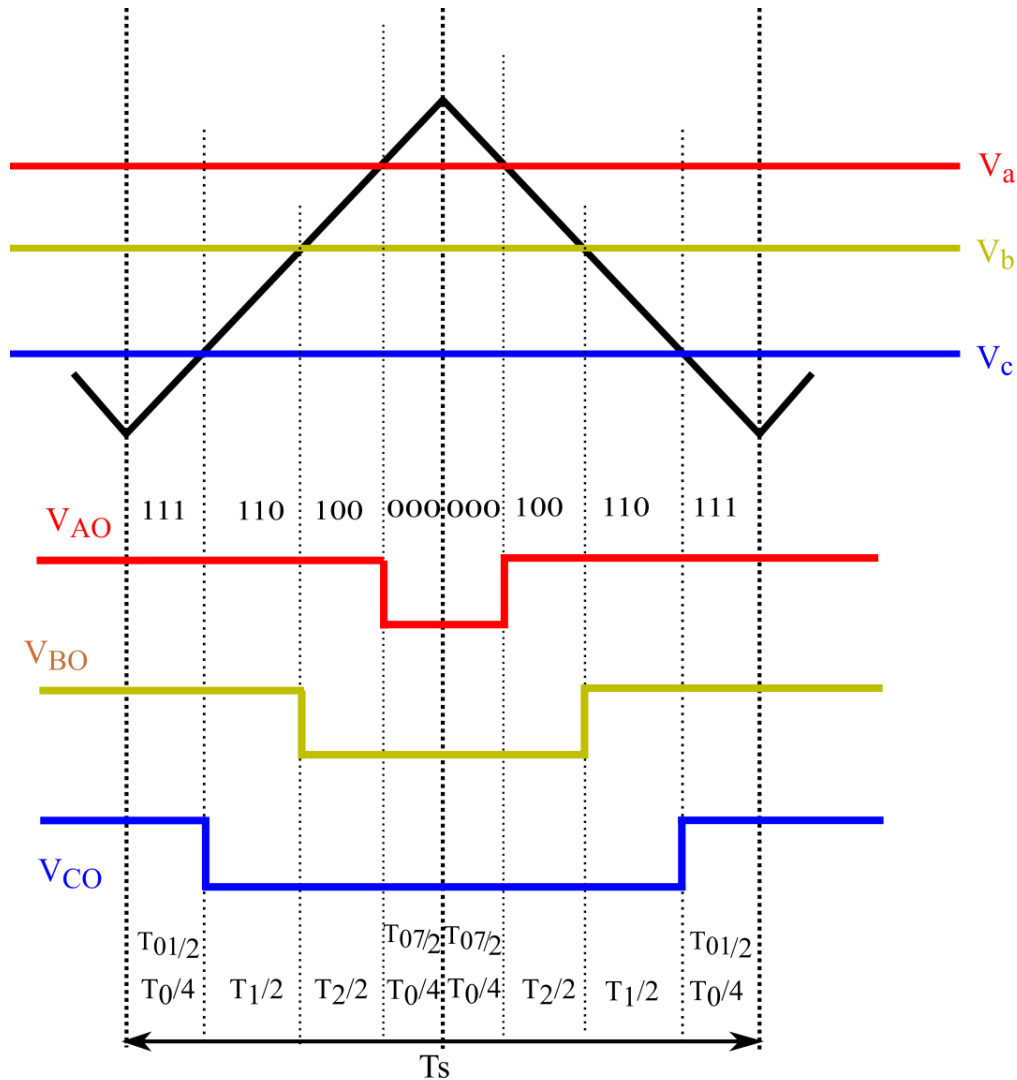
Zoomed view in one carrier in sine PWM



- Let us zoom further into one carrier.
- We observe that the two zero vector periods are not equal.
- In Sine PWM, T_{01} time period and T_{07} time period are not always equal.



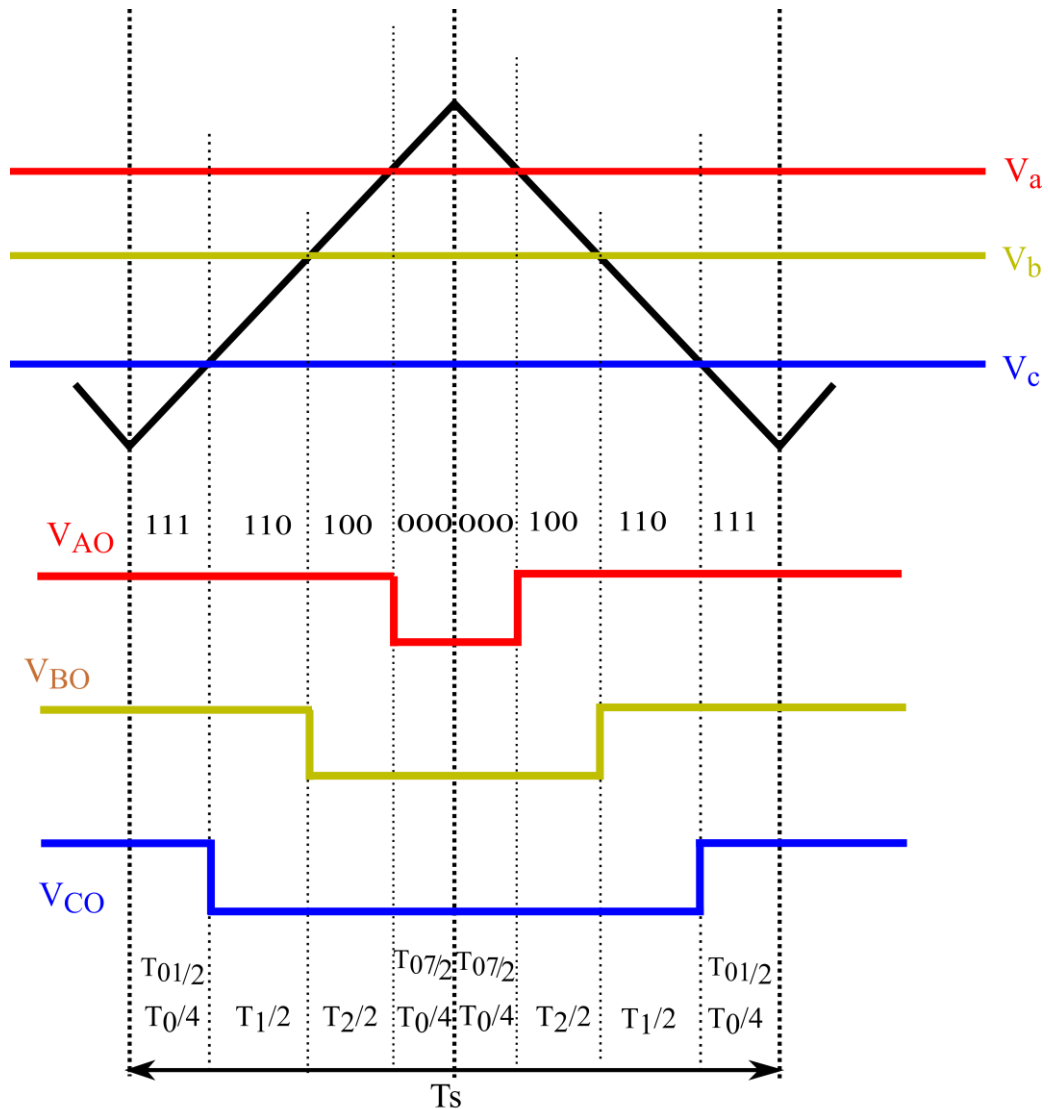
Switching sequence in space vector PWM



- The switching sequence here in one carrier period is 111-110-100-000-100-110-111.
- This is same as sine PWM, however the two zero periods are equal.



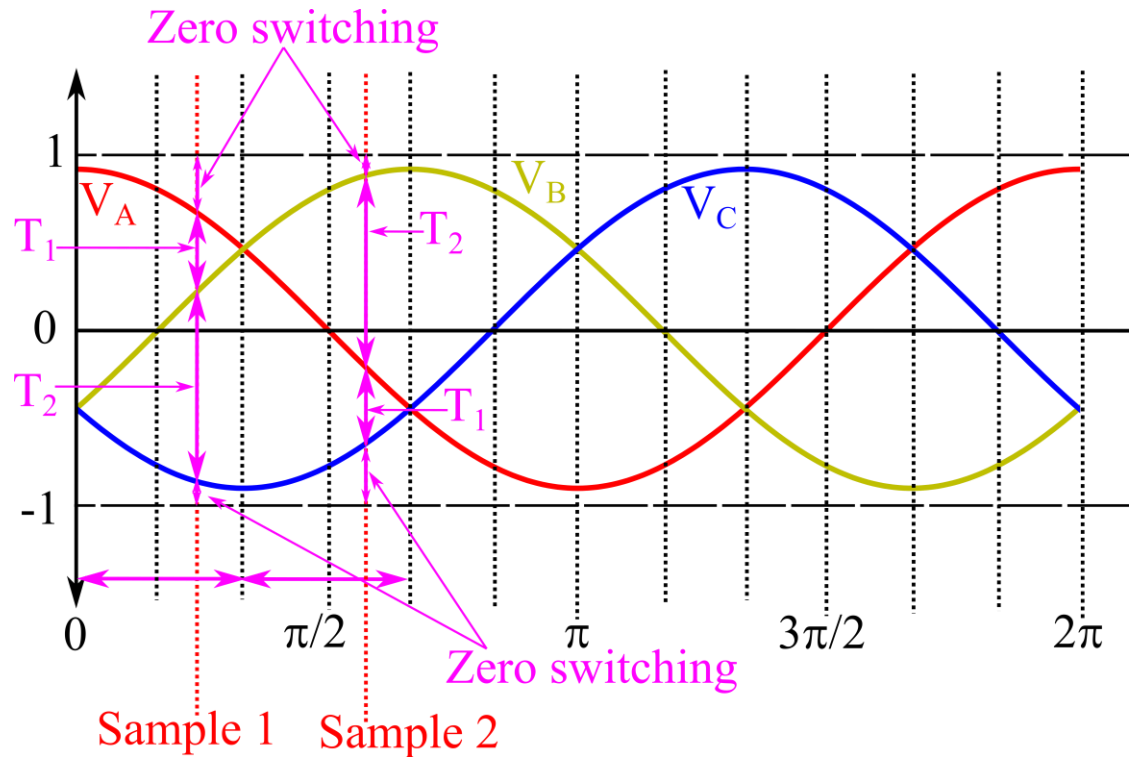
Switching sequence in space vector PWM



- Minimum switching is ensured.
- Switching frequency is same as carrier frequency.
- Thus space vector PWM is an extension of sine PWM, and can also be realized using carriers.



Extension of sine PWM



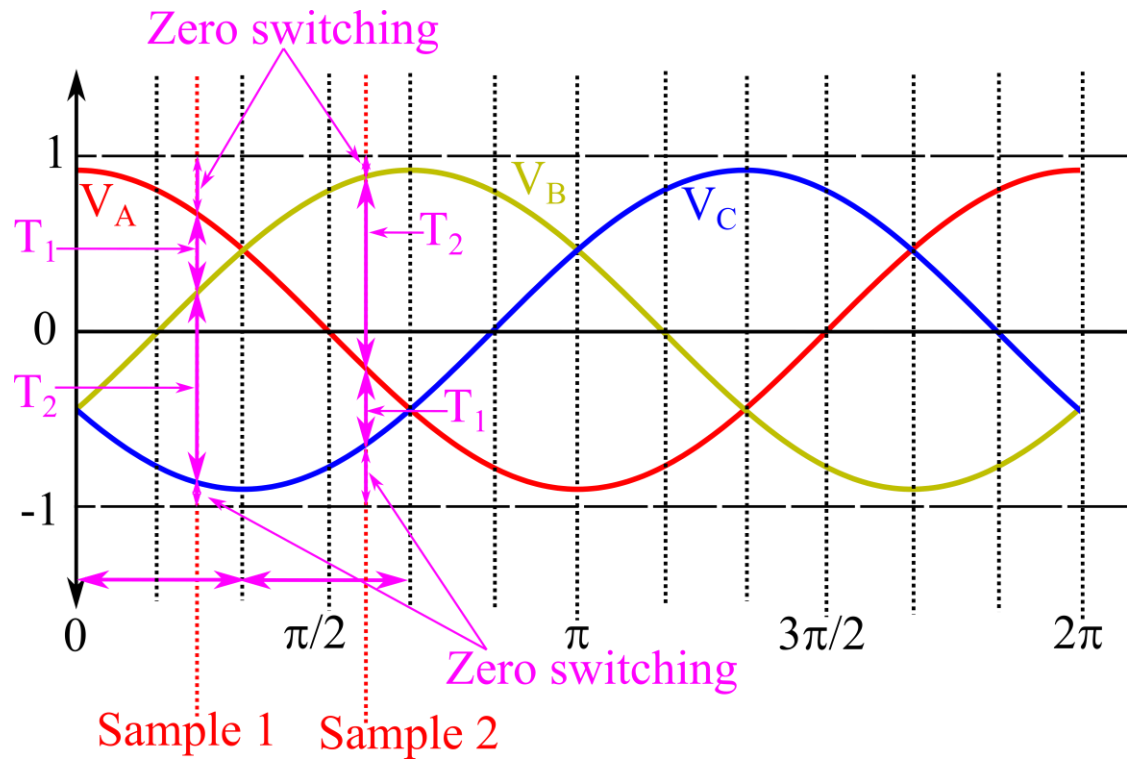
- The space vector PWM is an extension of sine PWM by addition of a common mode voltage.

$$\begin{aligned} v_a &= V_m \cos \theta, v_b = V_m \cos\left(\theta - \frac{2\pi}{3}\right), \\ v_c &= V_m \cos\left(\theta - \frac{4\pi}{3}\right) \end{aligned}$$

- What are the line voltages v_{ab} and v_{bc} ?
- $v_{ab} = \sqrt{3} V_m \sin\left(\frac{\pi}{3} - \theta\right)$
- $v_{bc} = \sqrt{3} V_m \sin \theta$



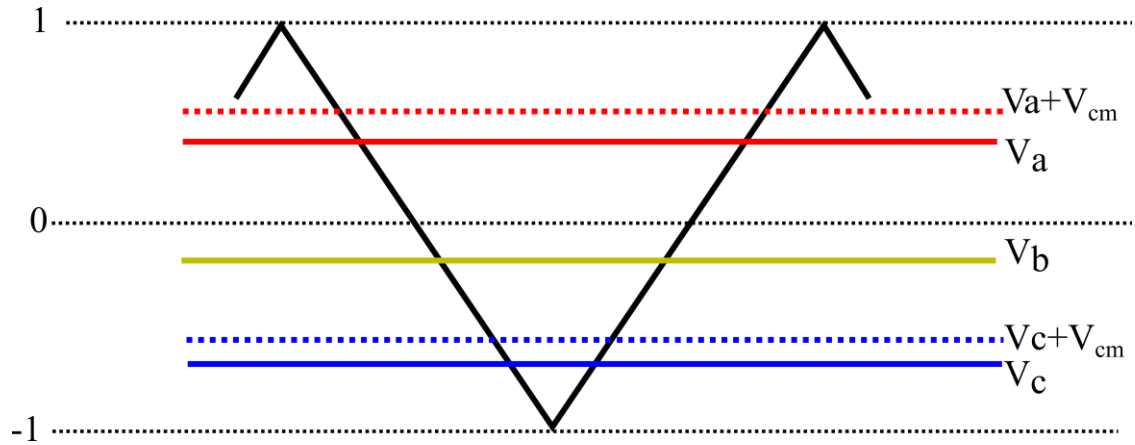
Extension of sine PWM



- The line voltage expressions follow the T_1 and T_2 expressions.
 - $v_{ab} = \sqrt{3} V_m \sin\left(\frac{\pi}{3} - \theta\right)$
 - $T_1 = \sqrt{3} \frac{V_R}{V_D} T_S \sin\left(\frac{\pi}{3} - \theta\right)$
 - $v_{bc} = \sqrt{3} V_m \sin \theta$
 - $T_2 = \sqrt{3} \frac{V_R}{V_D} T_S \sin \theta$
- The active vectors are represented by the line voltages.
- What about the zero vectors?



Common mode voltage

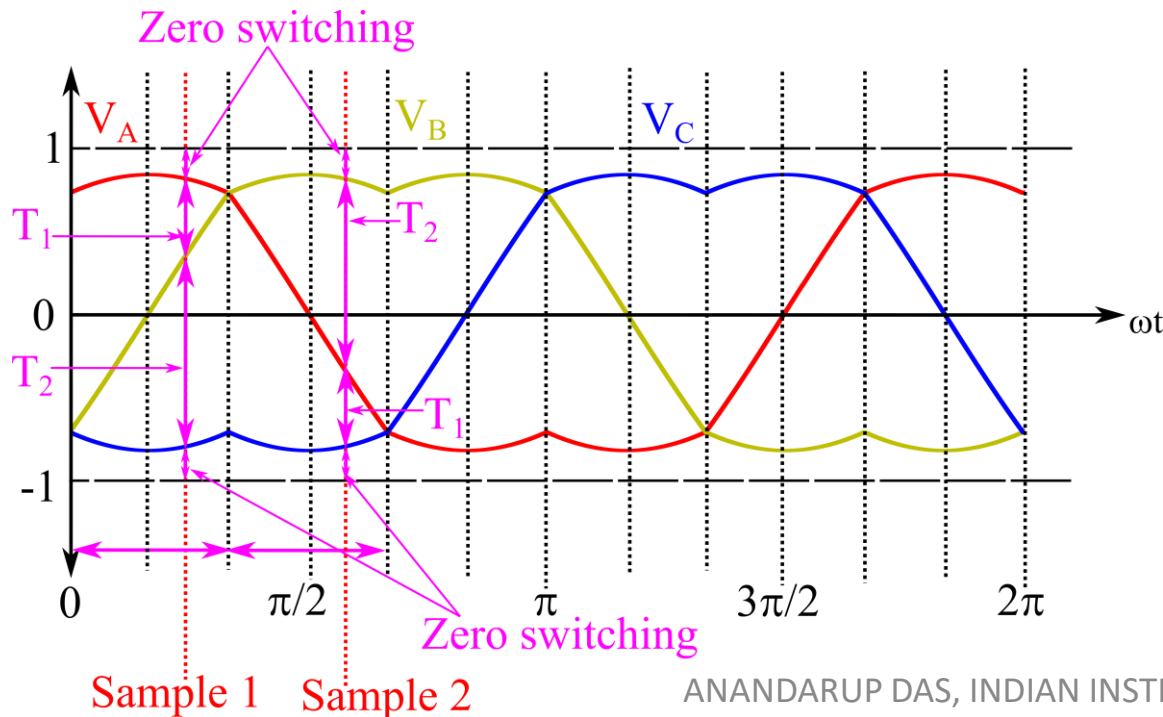


- To make the two zero vectors equal we have to add a common mode voltage.

- What should be its value?

- $1 - (v_a + v_{cm}) = 1 - v_c + v_{cm}$

- $v_{cm} = -\frac{v_a + v_c}{2}$

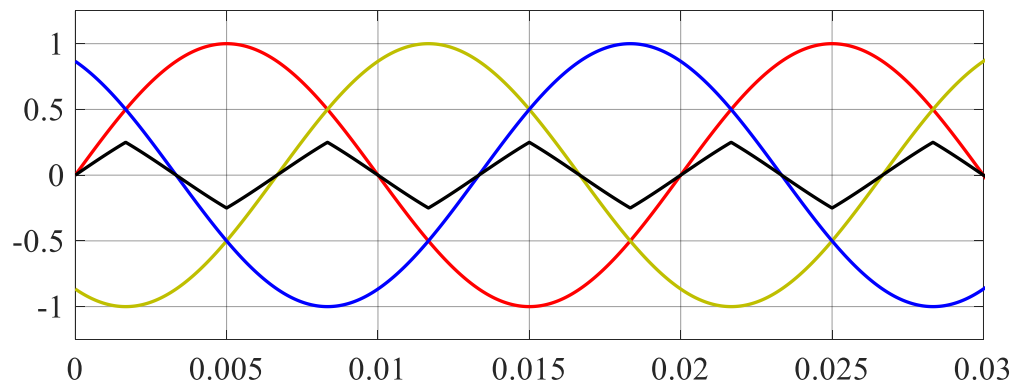


- In general, $v_{cm} = -\frac{v_{max} + v_{min}}{2}$

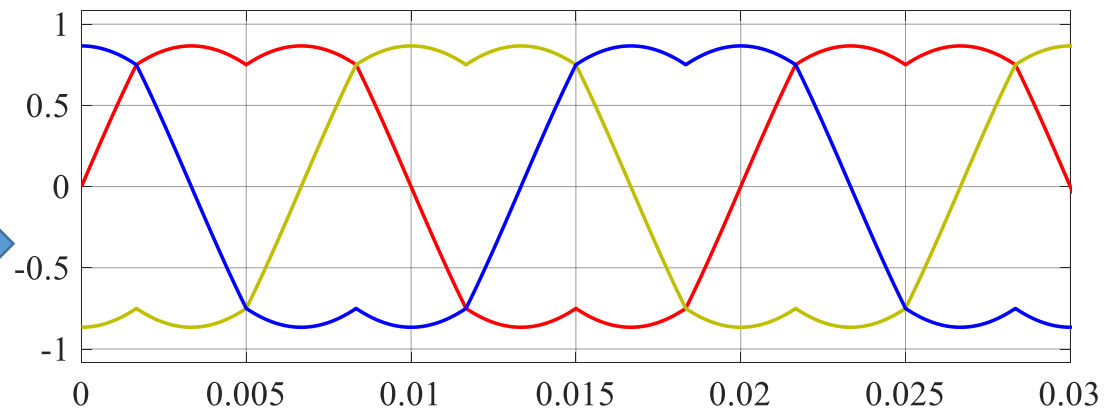


Waveforms

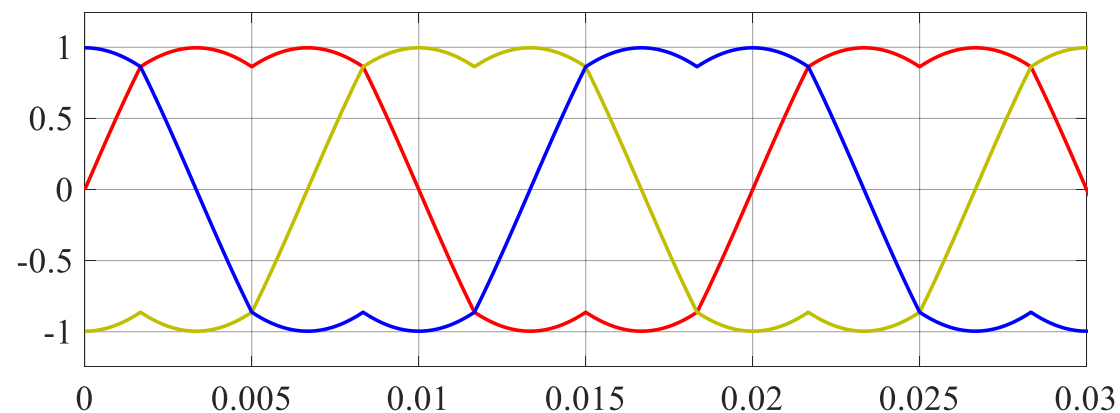
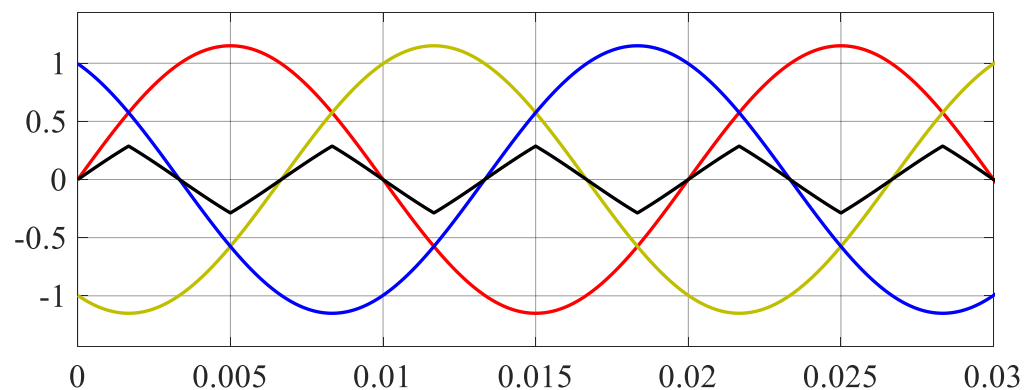
$m=1$



Resultant Waveform

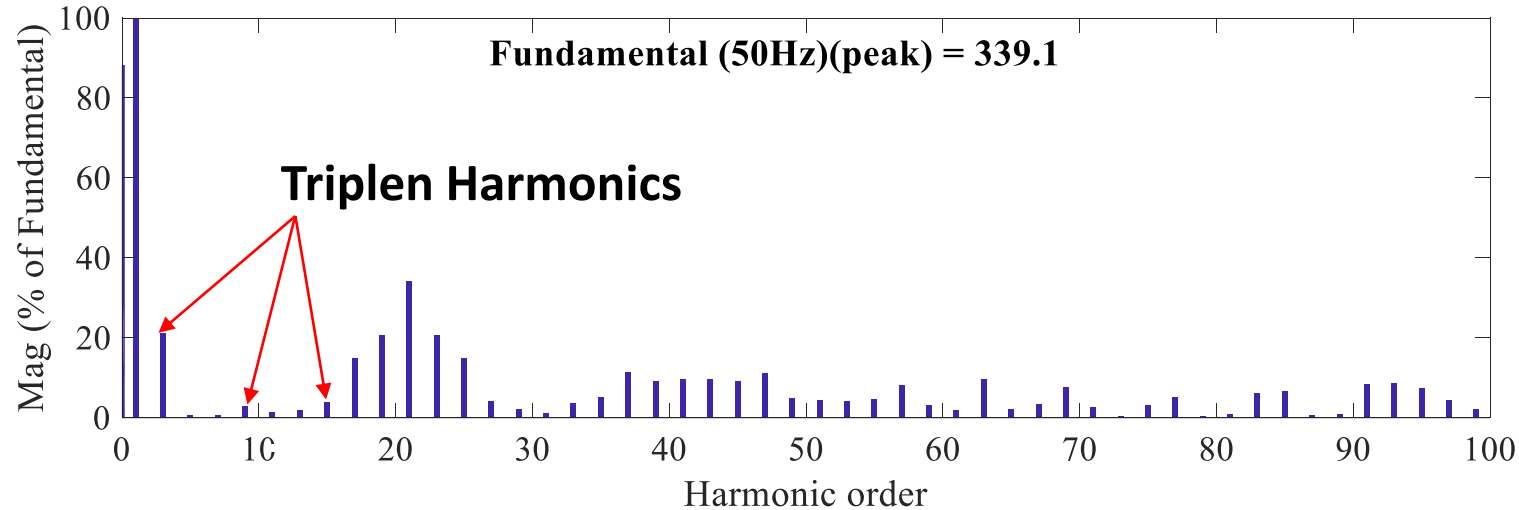
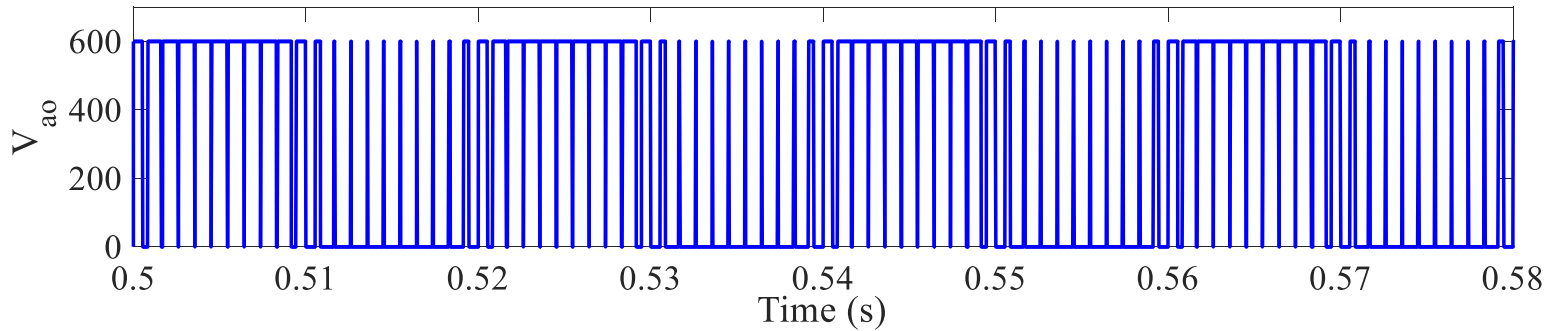


$m=1.15$



Simulation Waveforms

Pole Voltage

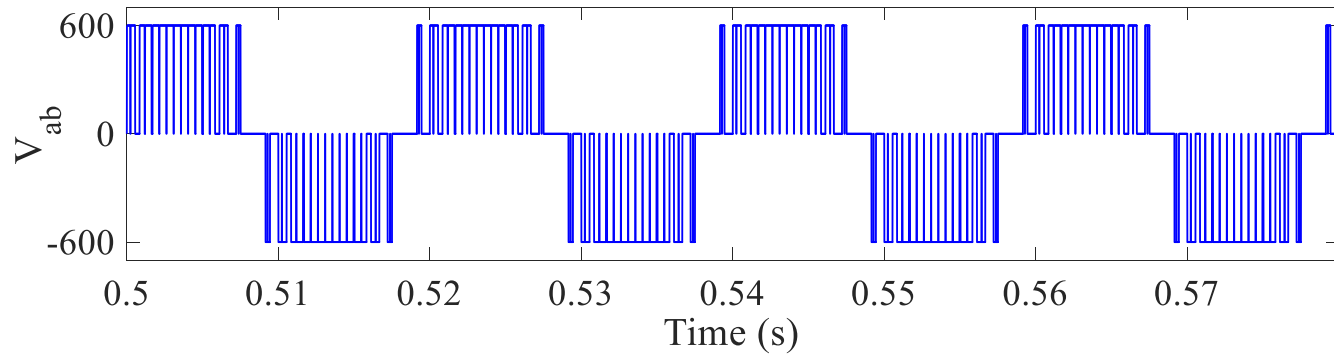


- $V_{DC}=600V$, the fundamental pole voltage is $(1.154*600)*0.5*0.98=339.27$ V
- $mf = 21$, harmonics reside around mf , $2mf$, $3mf$...

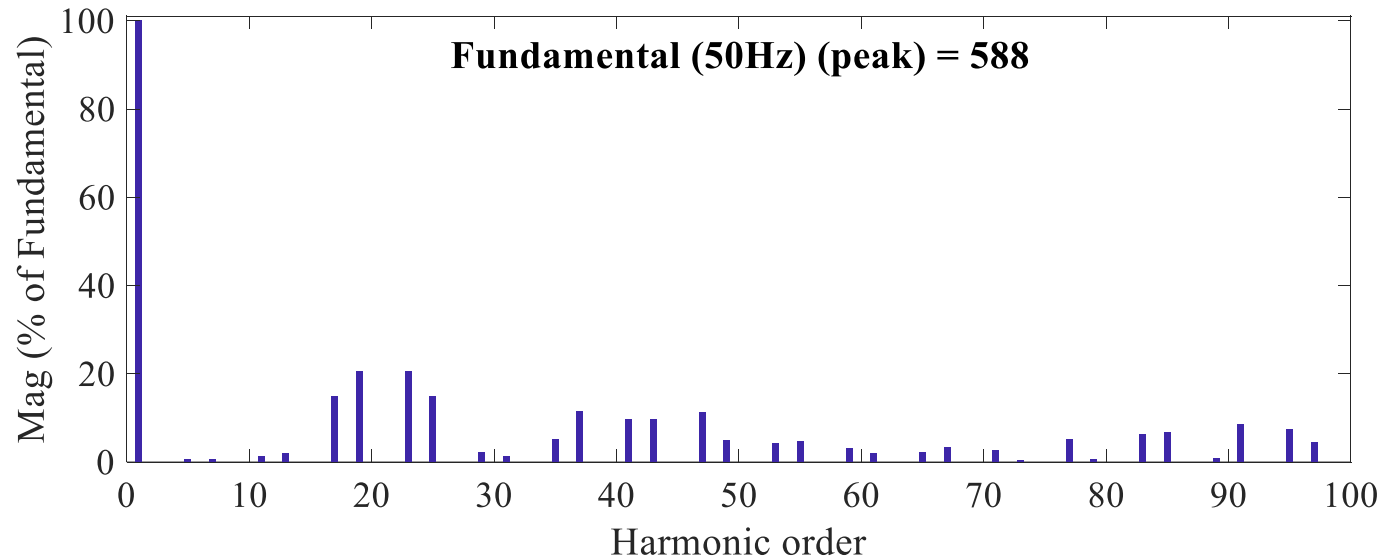


Simulation Waveforms

Line voltage

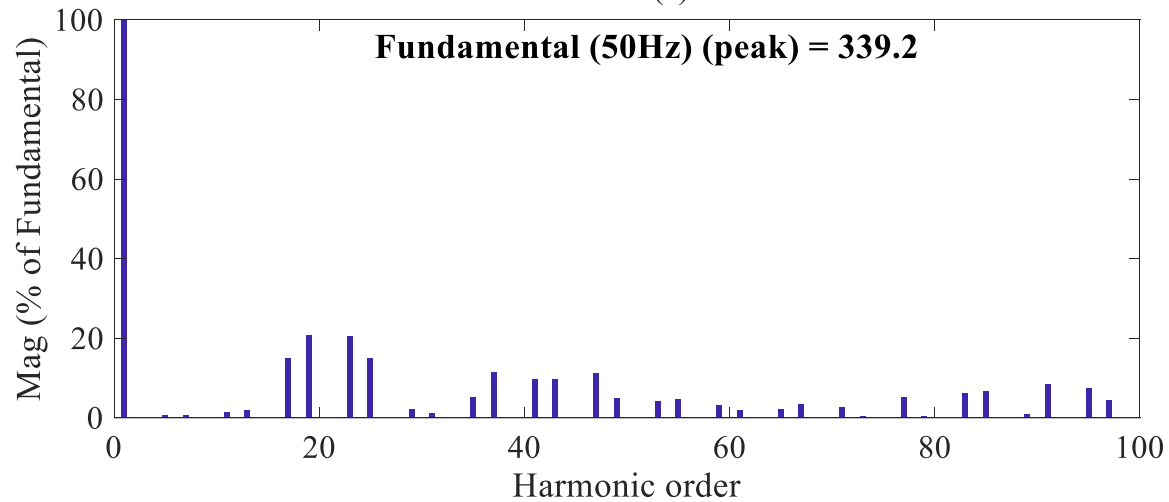
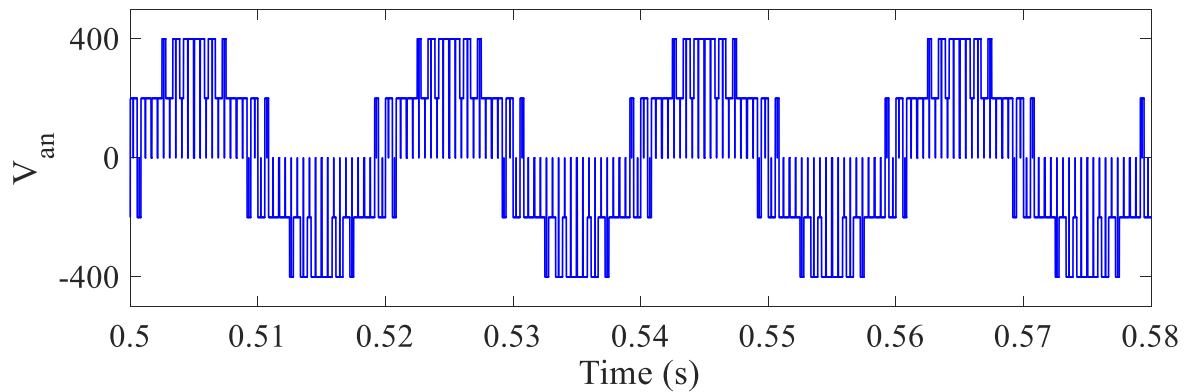


- $V_{DC}=600V$, the fundamental line voltage is $(1.154*0.98*600)*0.5*1.732*=587.62 V$
- $mf = 21$, harmonics reside around mf , $2mf$, $3mf \dots$



Simulation Waveforms

Pole voltage



- The phase voltage does not contain any triplen harmonic, so the phase current will be absent from it.

